# Circuit and System Analysis EHB 232E 

Prof. Dr. Müștak E. Yalçın<br>Istanbul Technical University Faculty of Electrical and Electronic Engineering<br>mustak.yalcin@itu.edu.tr

## Outline I

(1) Sinusoidal Steady-State Analysis-Cont.

- The Passive Circuit Elements in the Frequency Domain
- The Concept of Impedance and Admittance
- Resonance
- Phasor diagram


## The Passive Circuit Elements in the Frequency Domain

Resistors: From Ohm's law, if the current in a resistor varies sinusoidally with time, the voltage at the terminals of the resistor

$$
\begin{gathered}
v(t)=R \operatorname{Re}\left\{I_{R} e^{j w t}\right\} \\
v(t)=\operatorname{Re}\left\{R I_{R} e^{j w t}\right\} \\
\operatorname{Re}\left\{V_{R} e^{j w t}\right\}=\operatorname{Re}\left\{R I_{R} e^{j w t}\right\}
\end{gathered}
$$

from the properties of phasor

$$
V_{R}=R I_{R}
$$

or

$$
I_{G}=G V_{G}
$$

There is no phase shift between the current and voltage of resistor. The signals of voltage and current are said to be in phase.

## The Passive Circuit Elements in the Frequency Domain

Capacitor: Substituting the phasor representation of the current and phasor voltage at the terminals of a capacitor into $\left.i=C \frac{d v}{d t}\right)$

$$
\operatorname{Re}\left\{I_{C} e^{j w t}\right\}=C \frac{d \operatorname{Re}\left\{V_{C} e^{j w t}\right\}}{d t}
$$

using the properties of phasor

$$
\operatorname{Re}\left\{I_{C} e^{j w t}\right\}=\operatorname{Re}\left\{C V_{C} \frac{d e^{j w t}}{d t}\right\}=\operatorname{Re}\left\{C V_{C} j w e^{j w t}\right\}
$$

we get

$$
I_{C}=j w C V_{C}
$$

The current leads the voltage across the terminals of a capacitor by $90^{\circ}$.

## The Passive Circuit Elements in the Frequency Domain

## Inductor:

$$
V_{L}=j w L I_{L}
$$

The current lags the voltage by $90^{\circ}$. Independent current and voltage sources

$$
I_{k}=I_{m k} e^{j \theta_{k}}
$$

and

$$
V_{k}=V_{m k} e^{j \theta_{k}}
$$

## The Concept of Impedance and Admittance

The driving-point impedance of the one-port N (is formed by an arbitrary interconnection of linear time-invariant elements) at the frequency $w$ to be the ratio of the port-voltage phasor $V$ and the input-current phasor $I$ that is,

$$
Z(j w)=\frac{V}{l}
$$

Thus the amplitude of the port voltage is the product of the current amplitude times the magnitude of the impedance. $Z$ represents the impedence of the circuit element

$$
Z=\frac{V}{l}=R+j L
$$

$R$, is called resistance and $L$, is called reactance.
$Y$ represents the admittance of the circuit element

$$
Y=\frac{l}{V}=G+j B
$$

$G$, is called conductance and $B$, is called susceptance.

| Element | Impedance | Reactance | Admintance | Susceptance |
| :--- | :---: | :---: | :---: | :---: |
| Resistor | $R$ | - | $G$ | - |
| Capacitor | $-j / w C$ | $-1 / w C$ | $j w C$ | $w C$ |
| Inductor | $j w L$ | $w L$ | $-j / w L$ | $-1 / w L$ |

## Combining Impedance in Series and Parallel

Impedances in series can be combined into a single impedance by simply adding the individual impedances.

$$
Z=Z_{1}+Z_{2}+\ldots+Z_{n}
$$

when they in parallel

$$
Z=\left\{\frac{1}{Z_{1}}+\frac{1}{Z_{2}}+\ldots+\frac{1}{Z_{n}}\right\}^{-1}
$$

admittances in parallel:

$$
Y=Y_{1}+Y_{2}+\ldots+Y_{n}
$$

## Mutual Inductance



$$
\begin{gathered}
\phi_{1}=L_{1} i_{1}+M i_{2} \\
\phi_{2}=L_{2} i_{2}+M i_{1} \\
v_{1}=L_{1} \frac{d i_{1}}{d t}+M \frac{d i_{2}}{d t} \\
v_{2}=L_{2} \frac{d i_{2}}{d t}+M \frac{d i_{1}}{d t}
\end{gathered}
$$

## Example

Find the complete solution for $R=1 / 3 \Omega, C=1 F, L=1 / 2 H$, $V_{C}(0)=1 V$ ve $i_{L}(0)=1 A$ and $i(t)=\cos (w t)$.


$$
\begin{gathered}
Y_{e q}(j \omega)=\frac{1}{j \omega L}+\frac{1}{R}+j \omega C=\frac{R+j \omega L-\omega^{2} R L C}{j \omega R L} \\
V_{C}=Z_{e q}(j \omega) \cdot I_{K}=\frac{I_{K}}{Y_{e q}(j \omega)}=\frac{1}{Y_{e q}(j \omega)}=\frac{j \omega R L}{R-\omega^{2} R L C+j \omega L}
\end{gathered}
$$

Magnitude of $V_{C}(j \omega)$ is maximun when $\omega=\frac{1}{\sqrt{L C}}$ !


$$
Z\left(j \frac{1}{\sqrt{L C}}\right)=R
$$

## Resonance

Resonance occurs at a particular resonance frequency when the imaginary parts of impedances or admittances of circuit elements cancel each other.

## Example



Find the impedance $Z$ between the terminals:

$$
\begin{gathered}
Y=C j w+\frac{1}{L j w} \\
Z=R+\frac{L j w}{1-L C w^{2}} \\
Z=\frac{R-R L C w^{2}+L j w}{1-L C w^{2}}
\end{gathered}
$$

## Example



State equation for $R=1 / 5, C=1 F$ ve $L=1 / 6$

$$
\frac{d}{d t}\left[\begin{array}{c}
V_{C 1} \\
i_{L}
\end{array}\right]=\left[\begin{array}{cc}
-\frac{1}{R C} & -\frac{1}{C} \\
\frac{1}{L} & 0
\end{array}\right]\left[\begin{array}{c}
V_{C 1} \\
i_{L}
\end{array}\right]+\left[\begin{array}{c}
\frac{1}{R C} \\
0
\end{array}\right] e
$$

which is

$$
\frac{d}{d t}\left[\begin{array}{c}
V_{C 1} \\
i_{L}
\end{array}\right]=\left[\begin{array}{cc}
-5 & -1 \\
6 & 0
\end{array}\right]\left[\begin{array}{c}
V_{C 1} \\
i_{L}
\end{array}\right]+\left[\begin{array}{l}
5 \\
0
\end{array}\right] e
$$

The roots of the function

$$
\operatorname{det}\left\{\lambda I-\left[\begin{array}{cc}
-5 & -1 \\
6 & 0
\end{array}\right]\right\}=\lambda(\lambda+5)+6
$$

are the eigenvalues of $A$ which $\lambda_{1}=-3$ and $\lambda_{2}=-2$. Corresponding eigenvalues are $\left[\begin{array}{ll}1 & -2\end{array}\right]^{T}$ and $\left[\begin{array}{ll}1 & -3\end{array}\right]^{\top}$. Fundamental matrix

$$
M=\left[\begin{array}{ll}
e^{-3 t} & e^{-2 t} \\
-2 e^{-3 t} & -3 e^{-2 t}
\end{array}\right]
$$

The homogeneous solution

$$
x_{h}(t)=\left[\begin{array}{ll}
e^{-3 t} & e^{-2 t} \\
-2 e^{-3 t} & -3 e^{-2 t}
\end{array}\right]\left[\begin{array}{l}
\alpha_{1} \\
\alpha_{2}
\end{array}\right]
$$

The state transition matrix of the circuit

$$
\phi(t)=\left[\begin{array}{ll}
e^{-3 t} & e^{-2 t} \\
-2 e^{-3 t} & -3 e^{-2 t}
\end{array}\right]\left[\begin{array}{ll}
1 & 1 \\
-2 & -3
\end{array}\right]^{-1}
$$

## Example

Using the The Concept of Impedance, lets find the $V_{R}$

$$
\begin{aligned}
V_{R} & =R \frac{e}{Z} \\
& =R \frac{\left(1-L C w^{2}\right)}{L j w+R\left(1-L C w^{2}\right)} e
\end{aligned}
$$

Using the state-equation

$$
\begin{gathered}
V_{C}=\left[\begin{array}{ll}
1 & 0
\end{array}\right]\left[\begin{array}{cc}
j w+\frac{1}{R C} & \frac{1}{C} \\
-\frac{1}{L} & j w
\end{array}\right]^{-1}\left[\begin{array}{c}
\frac{1}{R C} \\
0
\end{array}\right] e \\
=\frac{\frac{j w}{R C}}{\frac{1}{L C}-w^{2}+\frac{j w}{R C}} \\
\begin{aligned}
V_{R}=e-V_{C} & =e\left\{1-\frac{\frac{j w}{R C}}{\frac{1}{L C}-w^{2}+\frac{j w}{R C}}\right\} \\
& =e\left\{\frac{\frac{1}{L C}-w^{2}}{\frac{1}{L C}-w^{2}+\frac{j w}{R C}}\right\} \\
& =\frac{R\left(1-L C w^{2}\right) e}{L j w+R\left(1-L C w^{2}\right)}
\end{aligned}
\end{gathered}
$$

## Example

```
Sekil 7.2
v1 1 0 sin(0 . 1 10) dc 0 ac 1
r 1 2 4k
l 2 0 2m
c 2 0 2m
.control
ac lin 1000.1 100
plot v(1,2)/4k
.endc
.end
```


## Example



YouTube Video: Example with two sources

## Phasor diagram

A phasor diagram shows the magnitude and phase angle of each phasor quantity in the complex number plane.


$$
V_{R}=V, V_{S}=V e^{-j 120^{\circ}}, V_{R S}=\sqrt{3} V e^{j 30^{\circ}}, V_{R S} ? f\left(V_{S}, V_{R}\right)
$$

