

# Basic of Electrical Circuits

## EHB 211E

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Lecture 6.

- 1 Circuit Elements
  - Two-terminal Elements
  - Two-port Elements
  - Multi-terminal Circuit Elements

# Two-terminal Elements

Two-terminal elements play a major role in electric circuits!

Two-terminal circuit elements are defined by the between basic variables which are current ( $i(t)$ ), voltage ( $v(t)$ ), charge ( $q(t)$ ) and flux ( $\phi(t)$ ). The units of them are Amperes, Volts, Coulomb and Weber, respectively.

Two pairs of the basic variables

$$i(t) = \frac{dq}{dt},$$

and

$$v(t) = \frac{d\phi}{dt},$$

are the definition.

# Two-terminal Elements

## Controlled circuit element (Dependent element)

If the relation between the terminal variable is given by the equation  $x = h(y, t)$ , this two-terminal element is called as a  $y$  controlled element e.g. voltage controlled voltage sources,...

## Time-invariant two-terminal element

A two-terminal element whose variables  $x$  and  $y$  fall on some fixed curve in the  $x - y$  plane at any time  $t$  is called a time-invariant circuit element e.g. Linear resistor  $Vv = Ri$ .

## $x - y$ characteristic

The curve on the  $x - y$  plane at any time  $t$  is called  $x - y$  characteristic e.g.  $v - i$  characteristic of linear resistor.

# Two-terminal Elements

## Bilateral property

A element has a  $x - y$  characteristics which is not symmetric with respect to the origin of the  $x - y$  plane.

## Linear element

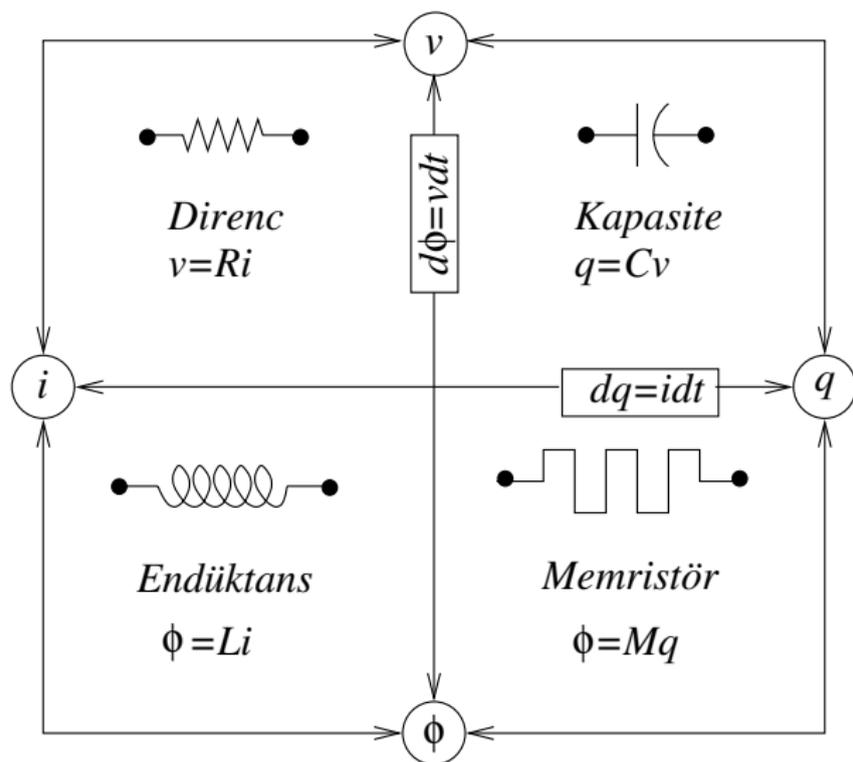
A linear element is an element with a linear relationship between its variables  $x$  and  $y$ .

## Linear

$f(x)$  is a function which satisfies the following two properties:

- Additivity (superposition):  $f(x + y) = f(x) + f(y)$ .
- Homogeneity :  $f(\alpha x) = \alpha f(x)$  for all  $\alpha$ .

# Basic circuit element diagram



A two-terminal element will be called a resistor if its voltage  $v$  and current  $i$  satisfy the following relation:

$$\mathbb{R} = \{(v, i) | f(v, i) = 0\}$$

This relation is called the  $v - i$  characteristic of the resistor and can be plotted graphically in the  $v - i$  plane. The equation  $f(v, i) = 0$  represents a curve in the  $v - i$  plane and specifies completely the two-terminal resistor.

The linear resistor is a special case of a resistor and satisfies Ohm's law which is

$$f(v, i) = v - Ri \quad \text{or} \quad f(v, i) = Gv - i$$

It means that the voltage across resistor is proportional to the current flowing through it.

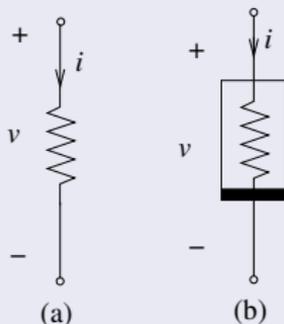
# Linear and Nonlinear Resistor

Ohm's law states

$$v = Ri \quad \text{or} \quad i = Gv$$

where the constant  $R$  is the resistance of the linear resistor measured in the unit of ohms ( $\Omega$ ), and  $G$  is the conductance measured in the unit of Siemens ( $S$ ). A resistor which is not linear is called nonlinear.

$$G = \frac{1}{R}$$

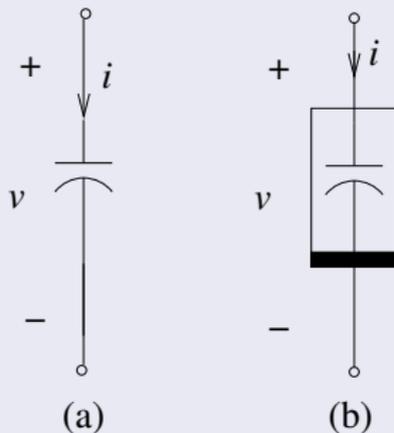


# Capacitor

A two-terminal element whose charge  $q(t)$  and voltage  $v(t)$  fall on some fixed curve in the  $q - v$  plane at any time  $t$  is called a time-invariant capacitor. Linear time-invariant capacitor is represented by the equations

$$q = Cv \text{ or } i = C \frac{dv}{dt}$$

Values of capacitors are specified in ranges of farads (F).



# Time-varying and Nonlinear Capacitor

If the  $q - v$  characteristic changes with time, the capacitor is said to be time-varying. Then the mathematical model becomes

$$q = C(t)v$$

and

$$i = \frac{dC}{dt}v + C(t)\frac{dv}{dt}$$

The most general case, a time-varying nonlinear capacitor is defined by a family of time-dependent and nonlinear  $q - v$  characteristics

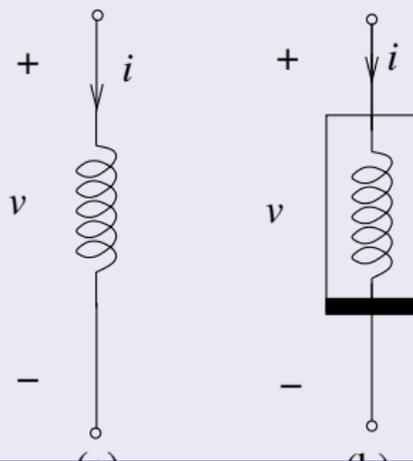
$$f(q, v, t) = 0$$

# Inductor

A two-terminal element whose flux  $\phi(t)$  and current  $i(t)$  fall on some fixed curve in the  $\phi - i$  plane at any time  $t$  is called a time-invariant inductor. The mathematical model of LTI inductor is

$$\phi = Li \text{ veya } v = L \frac{di}{dt}$$

Values of inductors are specified in ranges of Henry (H).



If the  $\phi - i$  characteristic changes with time, the inductor is said to be time-varying. Then the mathematical model becomes

$$v = L(t)i$$

and

$$v = \frac{dL}{dt}i + L(t)\frac{di}{dt}$$

The most general case, a time-varying nonlinear capacitor is defined by a family of time-dependent and nonlinear  $\phi - v$  characteristics

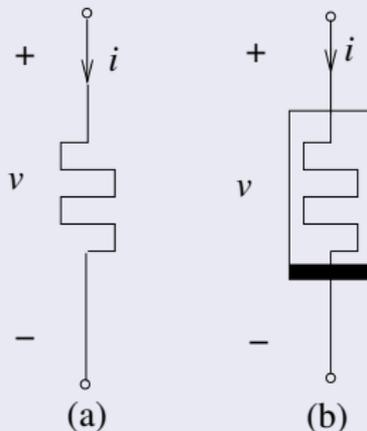
$$f(\phi, v, t) = 0$$

# Memristor

A two-terminal element whose flux  $\phi(t)$  and current  $q(t)$  fall on some fixed curve in the  $\phi - q$  plane at any time  $t$  is called a time-invariant memristor. The mathematical model of LTI memristor is

$$\phi = Mq.$$

Using the definitions, one can obtain  $v = Mi$  which is a ....



Nonlinear memristor is defined by a family of nonlinear  $\phi - v$  characteristics

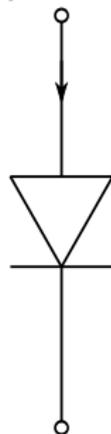
$$\phi = f(q)$$

It can be written in the form

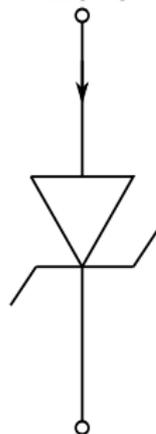
$$v = \frac{df(q)}{dq} i = \frac{df(q(t_0 + \int_{t_0}^t i(\tau) d\tau))}{dq} i$$

# Nonlinear Resistors

Diode



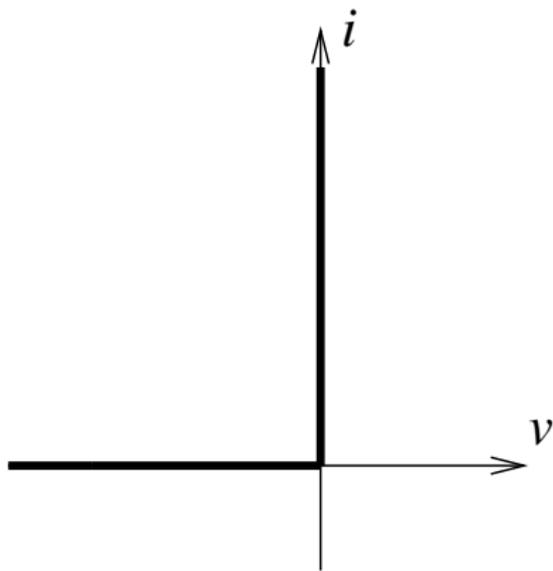
Zener diode



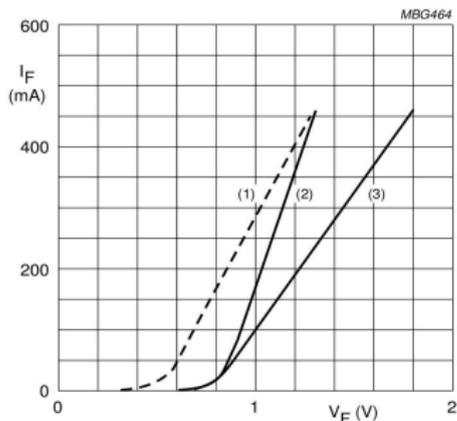
# Nonlinear Resistors: Diode

$$i = I_0 e^{(v/v_T - 1)}$$

where  $v_T = 0.026V$  ve  $I_0 \mu A$ .



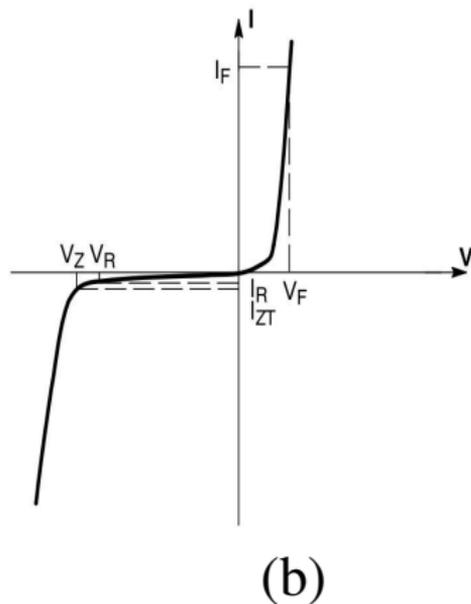
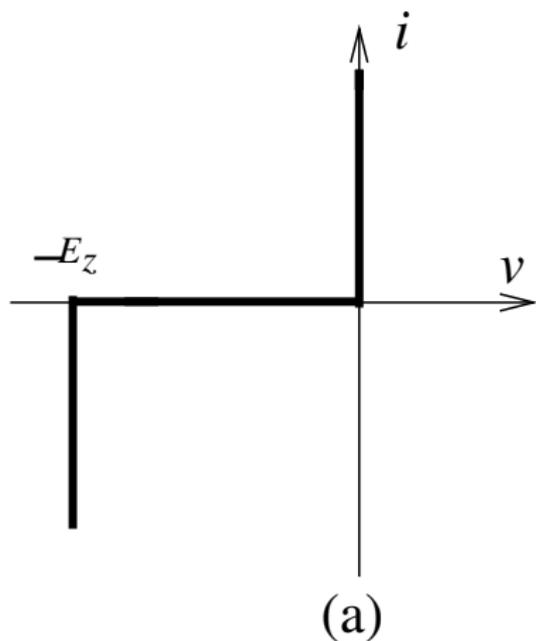
(a)



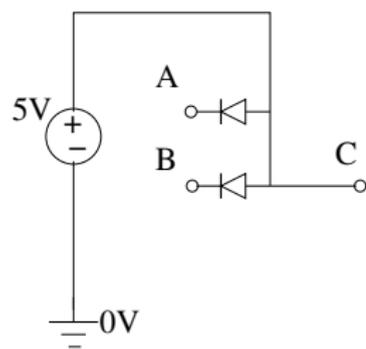
- (1)  $T_j = 175\text{ }^\circ\text{C}$ ; typical values.
- (2)  $T_j = 25\text{ }^\circ\text{C}$ ; typical values.
- (3)  $T_j = 25\text{ }^\circ\text{C}$ ; maximum values.

(b)

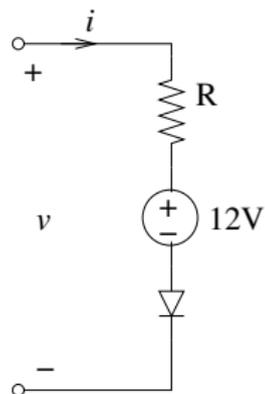
# Nonlinear Resistors: Zener diode



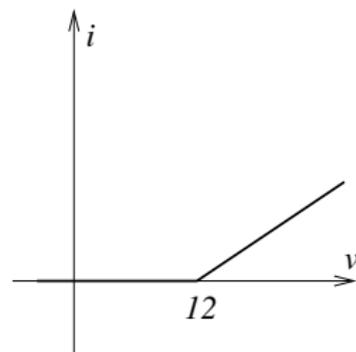
# Nonlinear Resistors



(a)



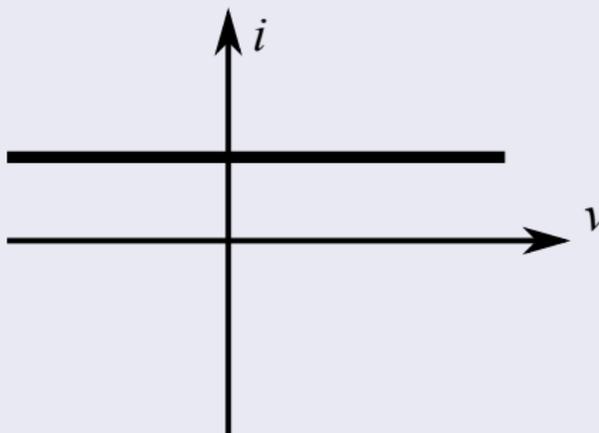
(b)



(c)

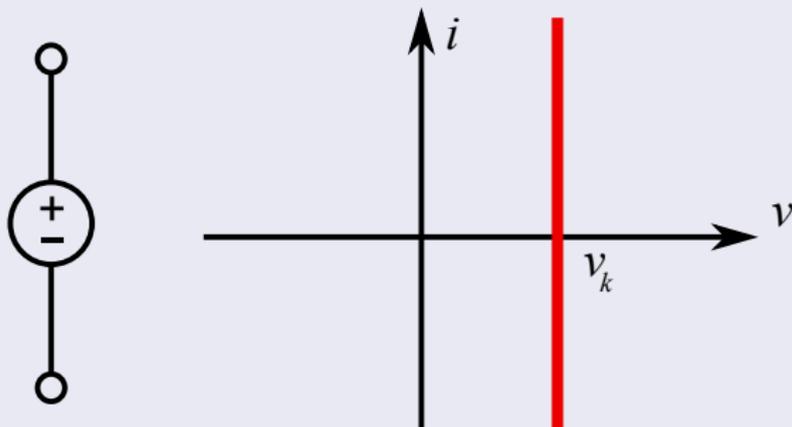
# Independent Sources

**Independent current source:** An independent current source is defined as a two-terminal circuit element whose current is a specified waveform  $i_s(\cdot)$  irrespective of the voltage across it.



# Independent Sources

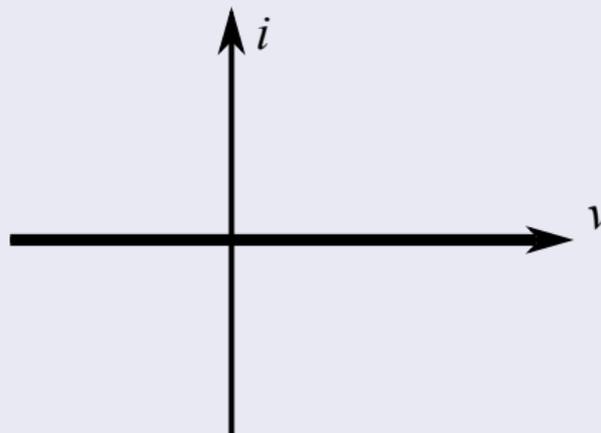
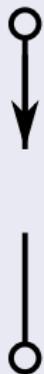
**Independent voltage source:** A two-terminal element is called an independent voltage source if the voltage across it is a given waveform  $v_s(\cdot)$  irrespective of the current flowing through it.



# Nonlinear resistor

Open circuit

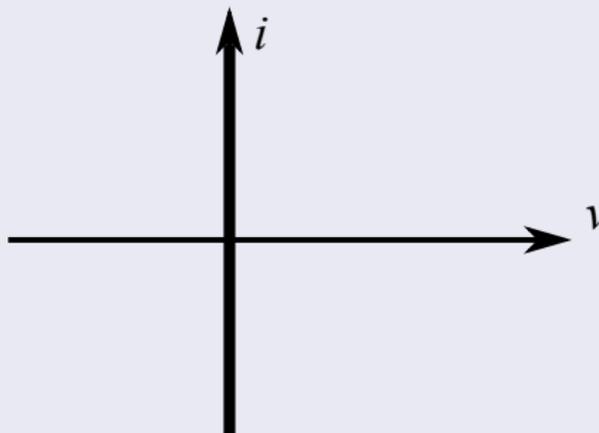
$$i(t) = 0$$



# Nonlinear resistor

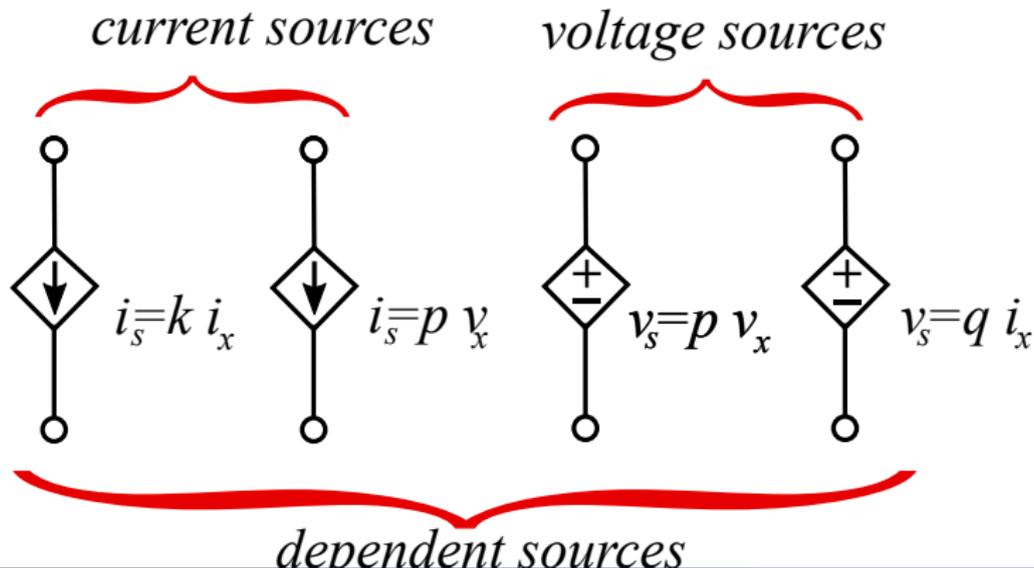
Short circuit

$$v(t) = 0$$



# Linear Controlled Sources

Dependent Source: A controlled source is a resistive two-port element consisting of two branches: a primary branch which is either an open circuit or a short circuit and a secondary branch which is either a voltage source or a current source. Diamond-shaped symbol to denote controlled sources.



# Linear Controlled Sources

## Current-Controlled Current Source (CCCS)

CCCS is characterized by two linear equations

$$\begin{bmatrix} v_1 \\ i_2 \end{bmatrix} = \begin{bmatrix} 0 & 0 \\ \alpha & 0 \end{bmatrix} \begin{bmatrix} i_1 \\ v_2 \end{bmatrix}$$

where  $\alpha$  is called the current transfer ratio.

## Current-Controlled Voltage Source (CCVS)

CCVS is characterized by two linear equations

$$\begin{bmatrix} v_1 \\ v_2 \end{bmatrix} = \begin{bmatrix} 0 & 0 \\ r & 0 \end{bmatrix} \begin{bmatrix} i_1 \\ i_2 \end{bmatrix}$$

where  $r$  is called the transresistance.

## Voltage-Controlled Current Source (VCCS)

$$\begin{bmatrix} i_1 \\ i_2 \end{bmatrix} = \begin{bmatrix} 0 & 0 \\ g & 0 \end{bmatrix} \begin{bmatrix} v_1 \\ v_2 \end{bmatrix}$$

where  $g$  is called the transconductance.

## Voltage-Controlled Voltage Source (VCVS)

$$\begin{bmatrix} i_1 \\ v_2 \end{bmatrix} = \begin{bmatrix} 0 & 0 \\ k & 0 \end{bmatrix} \begin{bmatrix} v_1 \\ i_2 \end{bmatrix}$$

where  $k$  is called the voltage transfer ratio.

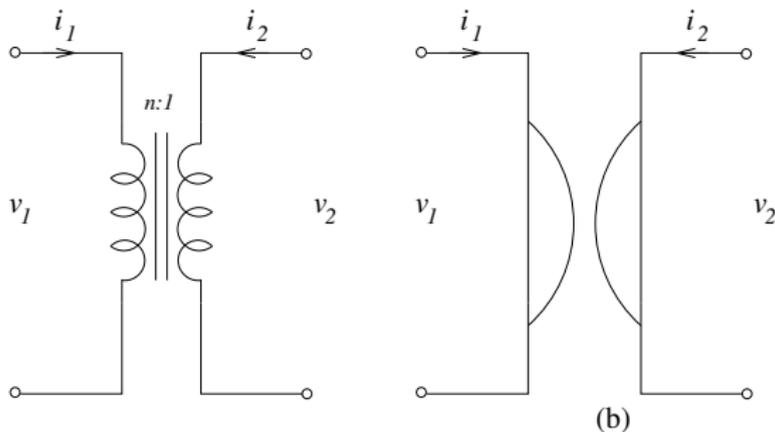
# Two-port Elements

## Two-port Elements

The ideal transformer is an ideal two-port resistive circuit element which is characterized by the following two equations:

$$\begin{bmatrix} v_1 \\ i_2 \end{bmatrix} = \begin{bmatrix} 0 & n \\ -n & 0 \end{bmatrix} \begin{bmatrix} i_1 \\ v_2 \end{bmatrix}$$

where  $n$  is a real number called the turns ratio.



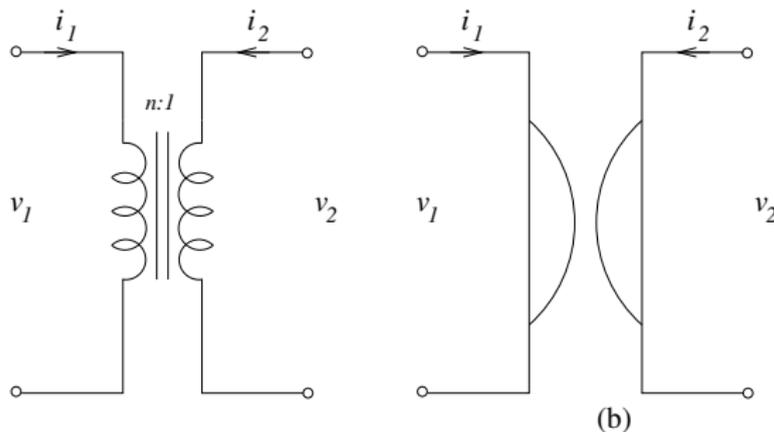
# Two-port Elements

## Jirator

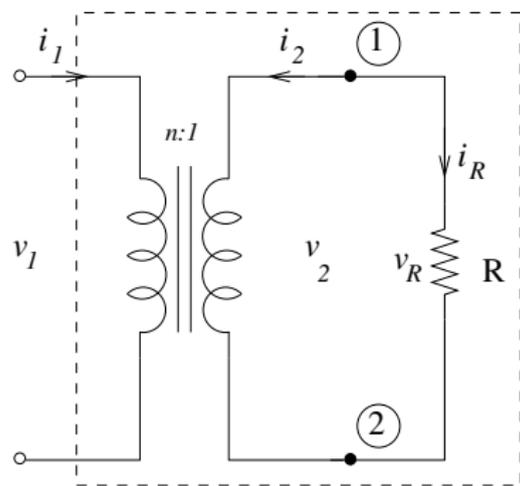
A gyrator is an ideal two-port element defined by the following equations:

$$\begin{bmatrix} i_1 \\ i_2 \end{bmatrix} = \begin{bmatrix} 0 & G \\ -G & 0 \end{bmatrix} \begin{bmatrix} v_1 \\ v_2 \end{bmatrix}$$

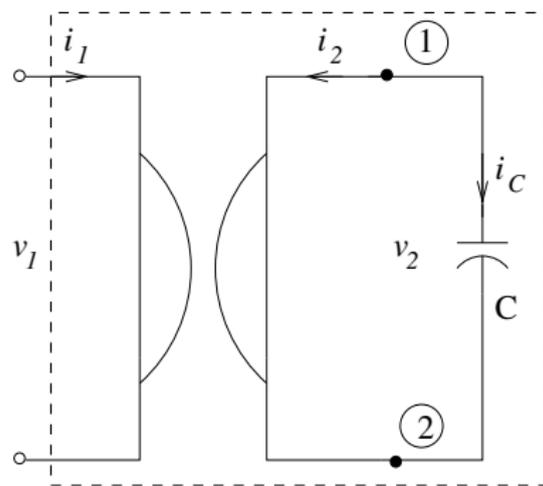
where the constant  $G$  is called the gyration conductance.



# Examples



(a)



(b)

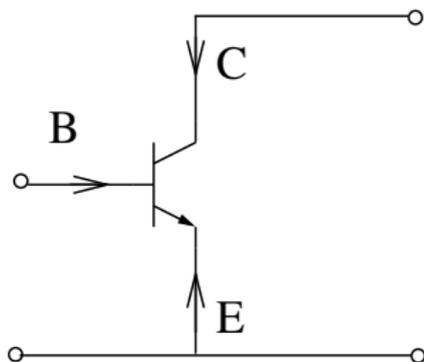
# Multi-terminal Circuit Elements

## npn- Bipolar Transistor

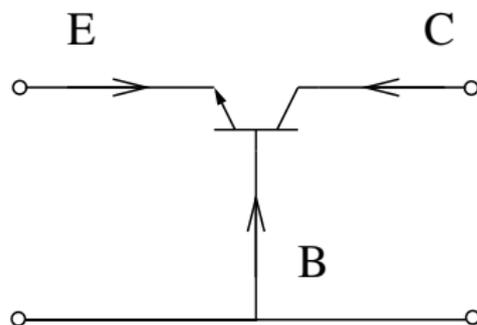
The most commonly used three-terminal device is the transistor. Low-frequency characterization is given by the one-dimensional diffusion model which yields the Ebers-Moll equations:

$$i_E = -I_{es}(e^{-v_{eb}/v_T} - 1) + \alpha_R I_{CS}(e^{-v_{cb}/v_T} - 1)$$

$$i_C = \alpha_F I_{es}(e^{-v_{eb}/v_T} - 1) + I_{CS}(e^{-v_{cb}/v_T} - 1).$$



(a)



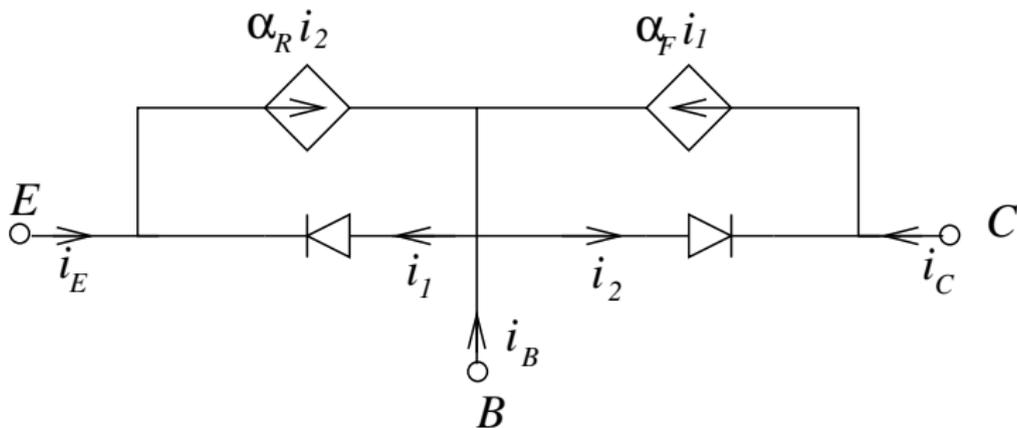
(b)

Modelling of the common emitter transistor configuration using diode and dependent sources. One can obtain

$$i_1 = I_{es}(e^{-v_{eb}/v_T} - 1)$$

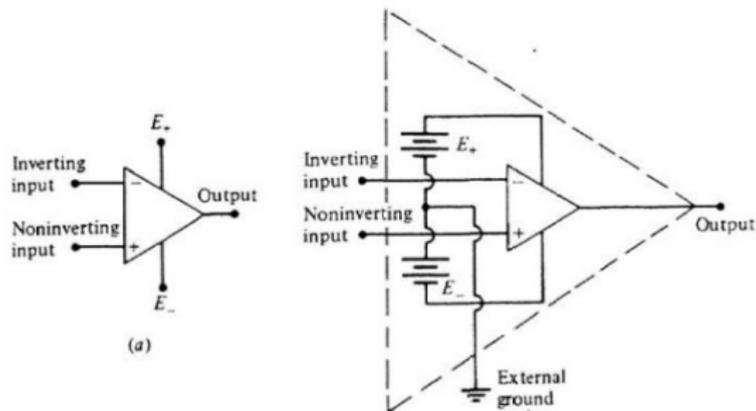
and

$$i_2 = I_{cs}(e^{-v_{cb}/v_T} - 1)$$



# Operational amplifier

Operational amplifiers (op amps) are multi-terminal devices. Terminals are labeled inverting input, non-inverting input, output,  $E_+$ ,  $E_-$  and external ground. A "biased" op amp can be considered as a 4-terminal device.



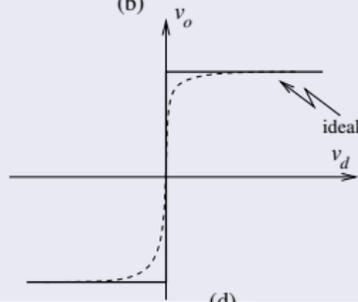
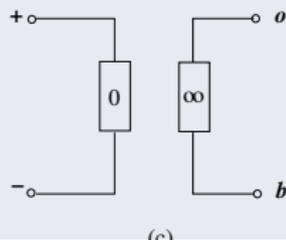
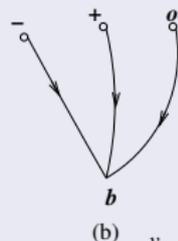
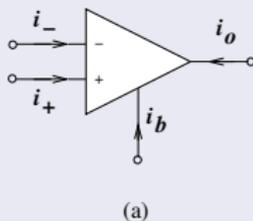
The variable  $v_d = v_+ - v_-$  is called the differential input voltage and will play an important role in op-amp circuit analysis.

## Operational amplifier

The op-amp terminal currents and voltages obey the following approximate relationships:

$$i_+ \approx 0, \quad i_- \approx 0 \quad \text{and} \quad v_o = G(v_+ - v_-)$$

where  $G$  called the open-loop voltage gain.  $v_o$  saturates at  $v_o = E_{\text{sat}}$  where  $E_{\text{sat}}$  is less than the power supply voltage.



# Operational amplifier

In a small interval  $-\epsilon < v_d < \epsilon$ , we have

$$G(v_+ - v_-) = v_o$$

which is called linear region.

In ideal op amp

$$i_+ = i_- = \epsilon = 0$$

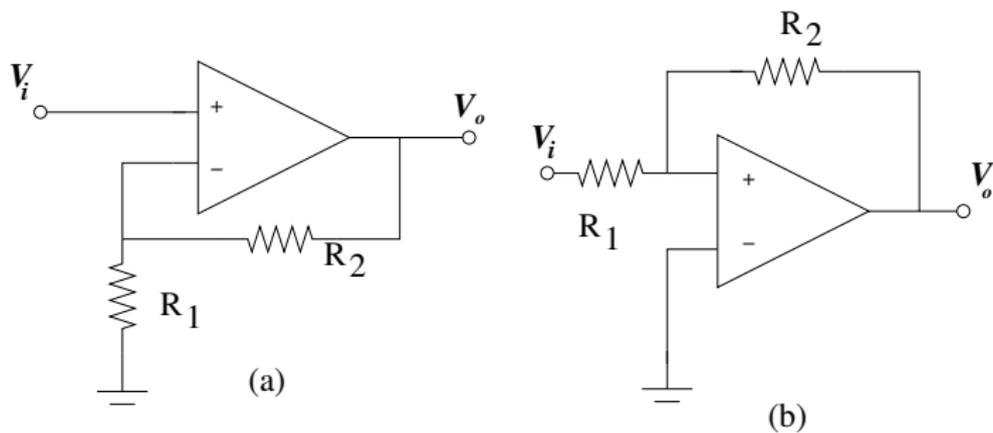
and

$$G = \infty$$

Using the gain formula, we will have

$$\begin{aligned}(v_+ - v_-) &= \frac{v_o}{G} \\(v_+ - v_-) &\approx 0\end{aligned}$$

# Examples



## Examples

Applying KCL at the node where the inverting input is connected and using the definition of op amp, we obtain

$$\begin{aligned} -G_1 V_- + G_2 (V_o - V_-) &= 0 \\ G(V_i + V_-) &= V_o \end{aligned}$$

Substituting the first eqn. into the second eqn.

$$V_o = \frac{G}{1 + KG} V_i$$

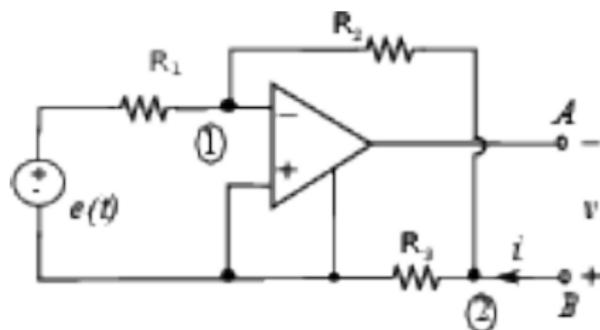
where  $K = \frac{R_1}{R_1 + R_2} \cdot \frac{G}{1 + KG}$  is the gain.  $G$  is too big therefore

$$V_o = \frac{1}{K} V_i$$

!

The same result can be obtained using the relation  $V_+ = V_-$ .

# Examples



What is the  $v - i$  characteristic ?

$$\frac{e}{R_1} = -\frac{V_{d2}}{R_2}$$

$$i_3 = -\frac{V_{d2}}{R_3} = -\frac{R_2 e}{R_3 R_1}$$

$$i = i_2 + i_3 = -\frac{e}{R_1} + \frac{R_2 e}{R_3 R_1} = -\frac{e}{R_1} \left(1 + \frac{R_2}{R_3}\right)$$